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# Communication Strategies and Coding for Relaying

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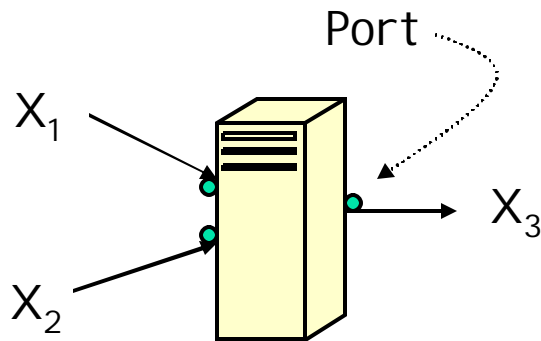
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# Outline

- 1) Network nodes
- 2) Wireline relaying
  - How data compression helps, and (later) why
- 3) Wireless relaying
  - Block-Markov coding
  - Half-duplex devices and timing modulation
  - Distributed codes

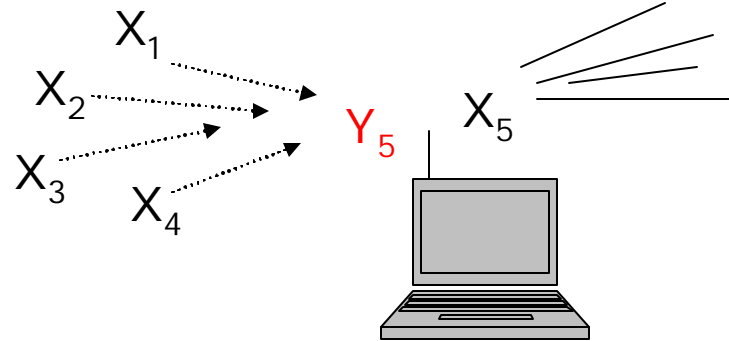
# 1) Network Nodes

## ■ Wireline



Node constraints:  
Suppose  $k$  ports can be active at once, e.g.,  $k=1$

## ■ Wireless

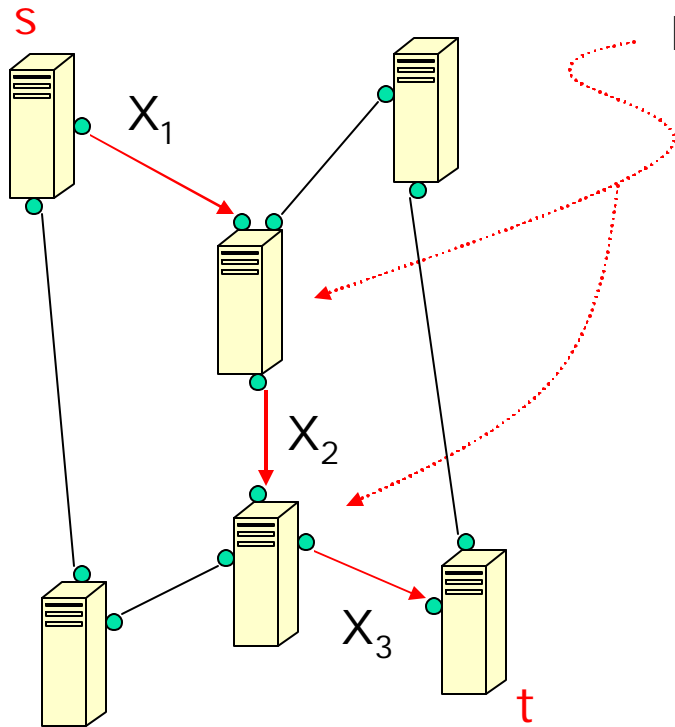


Half-duplex constraint:

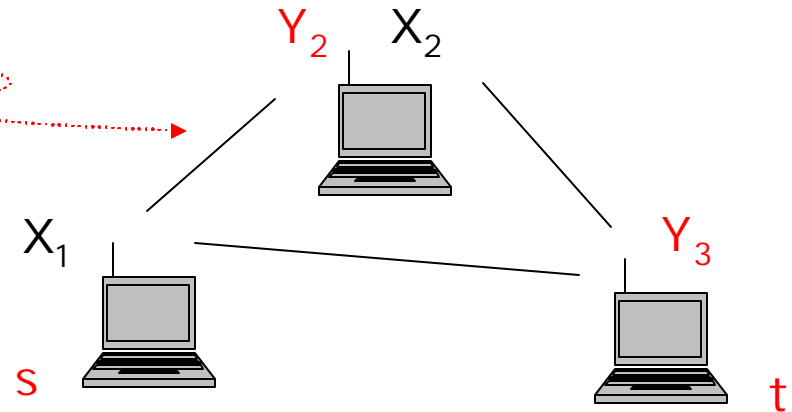
$$Y_t = \begin{cases} z_t + \sum_{s \neq t} \frac{h_{st}}{d_{st}^{\alpha/2}} X_s & \text{if } X_t = 0 \\ 0 & \text{else} \end{cases}$$

# Networks

## ■ Wireline



## ■ Wireless

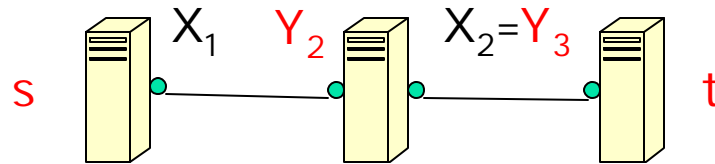


$$Y_2 = \begin{cases} Z_2 + \frac{h_{12}}{d_{12}^{a/2}} X_1 & \text{if } X_2 = 0 \\ 0 & \text{else} \end{cases}$$

$$Y_3 = Z_3 + \frac{h_{13}}{d_{13}^{a/2}} X_1 + \frac{h_{23}}{d_{23}^{a/2}} X_2$$

## 2) Wireline Relaying

- 3 node example:



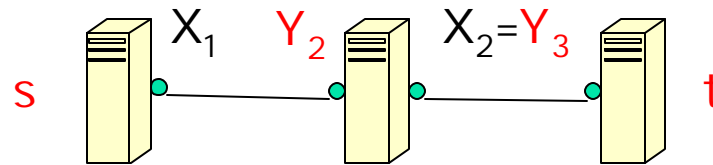
- Suppose 1 port/node can be active simultaneously.  
A link (channel) model:

$$\begin{aligned} \text{if } X_2=0 \text{ then } Y_2 &= X_1 \\ \text{if } X_2 \neq 0 \text{ then } Y_2 &= 0 \end{aligned}$$

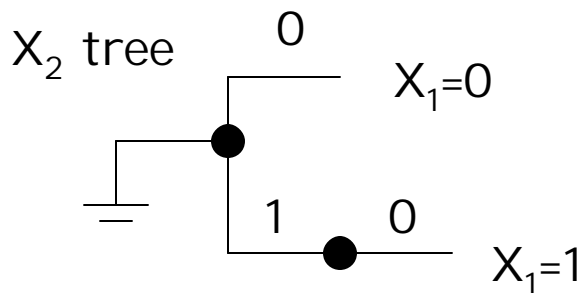
- Suppose the random variables are bits\*.

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\*Usually the  $X_t$  are **packets**, and not bits, but the following gives the general idea



- Guess: capacity is  $\frac{1}{2}$  bit/use (or packet/use) ?
- A “decompression” code at node 2:

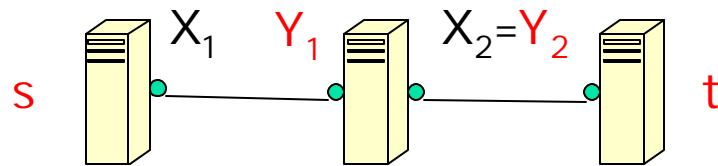


Node 2 transmits appropriate branch labels upon receiving  $X_1$ .  
For example:

$X_1 = 0, 1, X0, 0, 1, X1, X1, X0$

$X_2 = 0, 0, 10, 0, 0, 10, 10, 10, 0$

- 1st network edge: every  $X_2$  word has one zero
- 2nd network edge:  $R = 1/E[L_2] = 2/3$  bits/use !

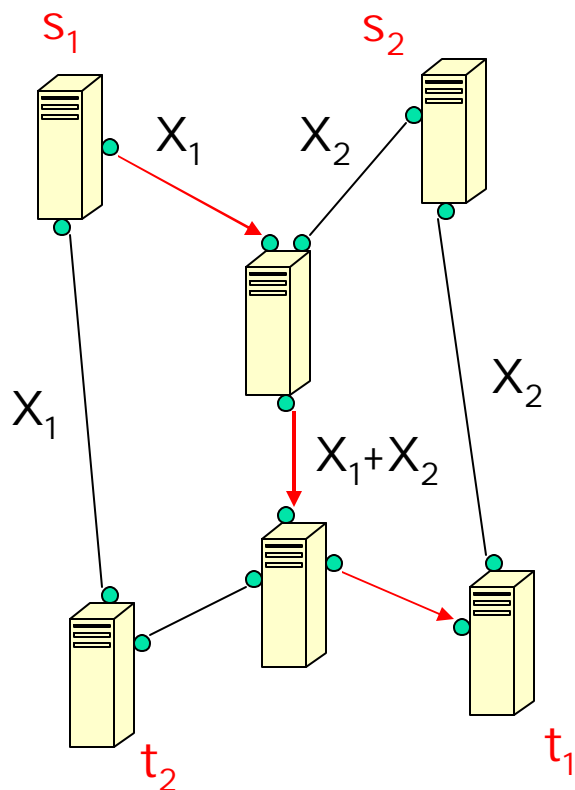


- Better compression codes (e.g., Huffman codes, arithmetic source codes) achieve  $R = 0.773$  bits/use with  $\Pr[X_2=0] = 0.773^*$ .
- How can we understand this gain?  
Is 0.773 the capacity of this network?

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\*This is when  $\Pr[X_2=0] = h(\Pr[X_2=0])$ , where  $h(x) = -x\log_2 x - (1-x)\log_2(1-x)$  is Shannon's binary entropy function

- Network implications:



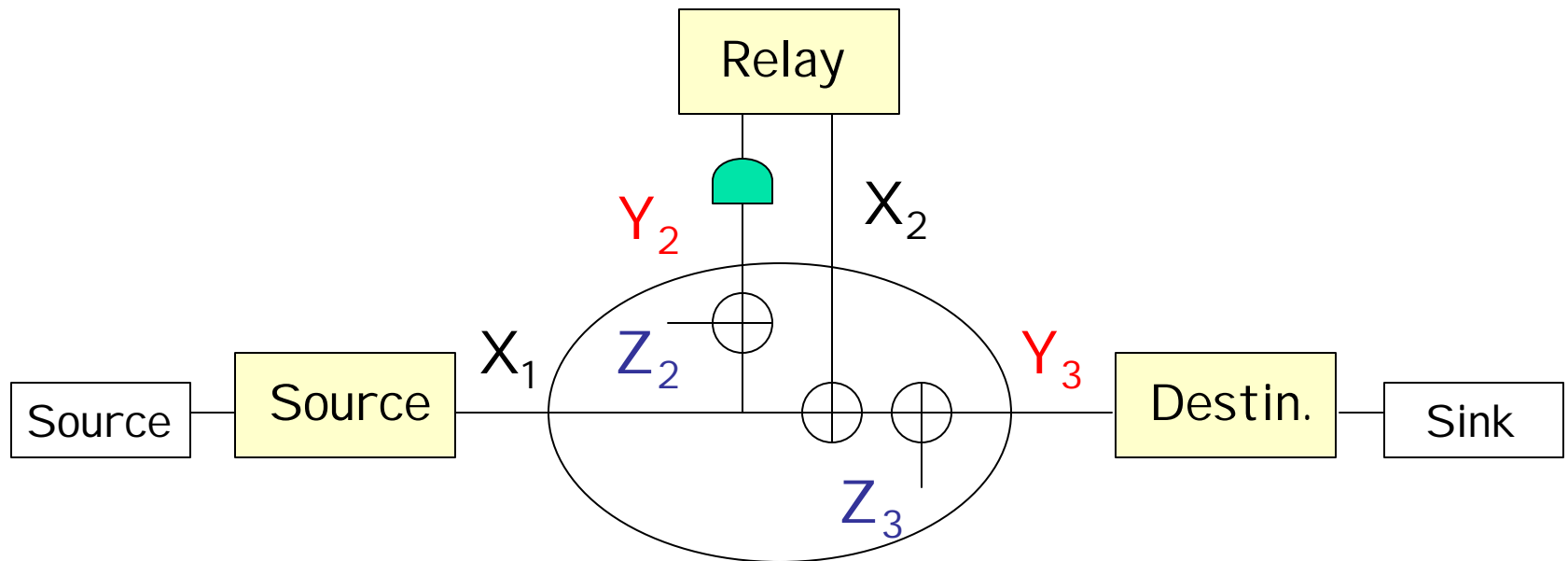
- Suppose every node has a 1-port constraint
- Basic routing throughput: 1/2 bit/use
- Basic network coding: 2/3 bits/use
- **Relay** routing: 0.732 bits/use\*

\*in general, one should combine network coding and relaying

\*for packets, the gains are smaller but do permit "covert" communication



# 3) Wireless Relaying

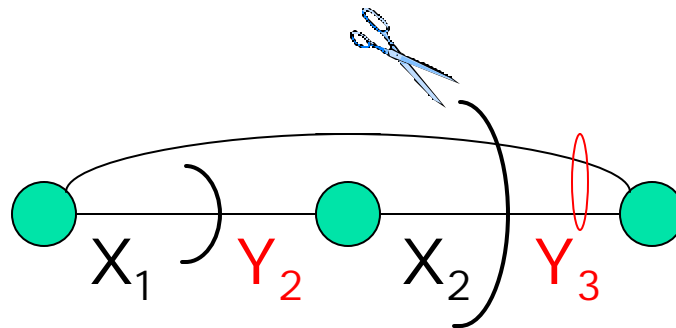


- Complex alphabets, lossless paths, full duplex (for now)
- Power:  $E[|X_{ti}|^2] = P_t$ , for  $t=1,2$ , all  $i$ , Gaussian noise  $Z_t$  (var. 1)
- No relay ( $X_2=0$ ): the capacity is  $\max_{P(X_1)} I(X_1; Y_3) = \log(1+P_1)$
- The capacity of the above problem is still open !

# Coding Methods

- Various relaying strategies exist:
  - Amplify-and-forward (amplify  $Y_2$ )
  - Decode-and-forward (includes basic multi-hopping)
  - Compress-and-forward (quantize  $Y_2$  and encode)
- The best decode-and-forward scheme achieves

$$R = \max_{P(x_1, x_2)} \min [ I(X_1; Y_2 | X_2), I(X_1 X_2; Y_3) ]$$



# Block-Markov Coding

- Carleial's decode-and-forward strategy (1982):  
choose  $P_1' = P_1$  and set  $\beta = [(P_1 - P_1')/P_2]^{1/2}$

	Block 1	Block 2	Block 3	Block 4
$\underline{x}_1$	$\underline{x}'_1(W_1) + \beta \underline{x}_2(1)$	$\underline{x}'_1(W_2) + \beta \underline{x}_2(W_1)$	$\underline{x}'_1(W_3) + \beta \underline{x}_2(W_2)$	$\underline{x}'_1(1) + \beta \underline{x}_2(W_3)$
$\underline{x}_2$	$\underline{x}_2(1)$	$\underline{x}_2(W_1)$	$\underline{x}_2(W_2)$	$\underline{x}_2(W_3)$

- Relay:  $R < I(X_1; Y_2 | X_2) = \log(1 + P_1')$
- Dest:  $R < I(X_1 X_2; Y_3) = \log(1 + P_1' + (1 + \beta)^2 P_2)$
- 3 additions to basic multi-hopping: Tx at same time, coherent combining (sync!), Rx with all information

# Fading Channels (Random Phases)

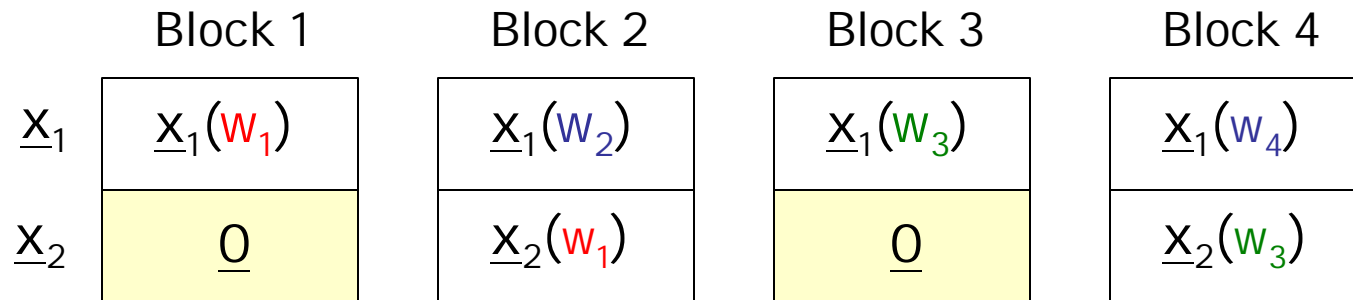
- Phase sync. is often not possible and  $\beta=0$  is best

	Block 1	Block 2	Block 3	Block 4
$\underline{x}_1$	$\underline{x}_1(w_1)$	$\underline{x}_1(w_2)$	$\underline{x}_1(w_3)$	$\underline{x}_1(1)$
$\underline{x}_2$	$\underline{x}_2(1)$	$\underline{x}_2(w_1)$	$\underline{x}_2(w_2)$	$\underline{x}_2(w_3)$

- A recent result (KGG, 2003): DF achieves capacity if the relay is near, but not necessarily co-located with, the source (for the **full-duplex** case)

# Half-Duplex Devices

- A natural approach (not DF in a strict sense):



- But we can do better: we can modulate the listen/talk times. Let  $M_2=0$  or  $1$  if the relay listens or talks, resp. We can achieve:

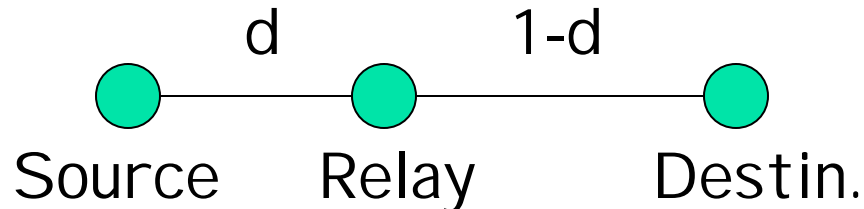
$$\begin{aligned}
 R &= \min [ I(X_1; Y_2 | X_2 M_2), I(X_1 X_2 M_2; Y_3) ] \\
 &= \min [ I(X_1; Y_2 | X_2 M_2), I(M_2; Y_3) + I(X_1 X_2; Y_3 | M_2) ]
 \end{aligned}$$

# Timing Modulation

- The above explains why our wireline rate improved!
  - In that example, we had  $M_2 = X_2$  and
$$I(X_1; Y_2 | X_2 M_2) = I(M_2; Y_3) = 0.773$$
$$I(X_1 X_2; Y_3 | M_2) = 0$$
- Notes:
  - Recent result (K, 2004): DF **with** timing modulation achieves capacity if the relay is near, but not nec. co-located with, the source (**and** if duplex ratio/mode power/mod. is limited)
  - Timing modulation improves AF, DF, CF rates
  - If one cannot modulate the timing every symbol, then (d,k) constraints become interesting

# A Geometric Example

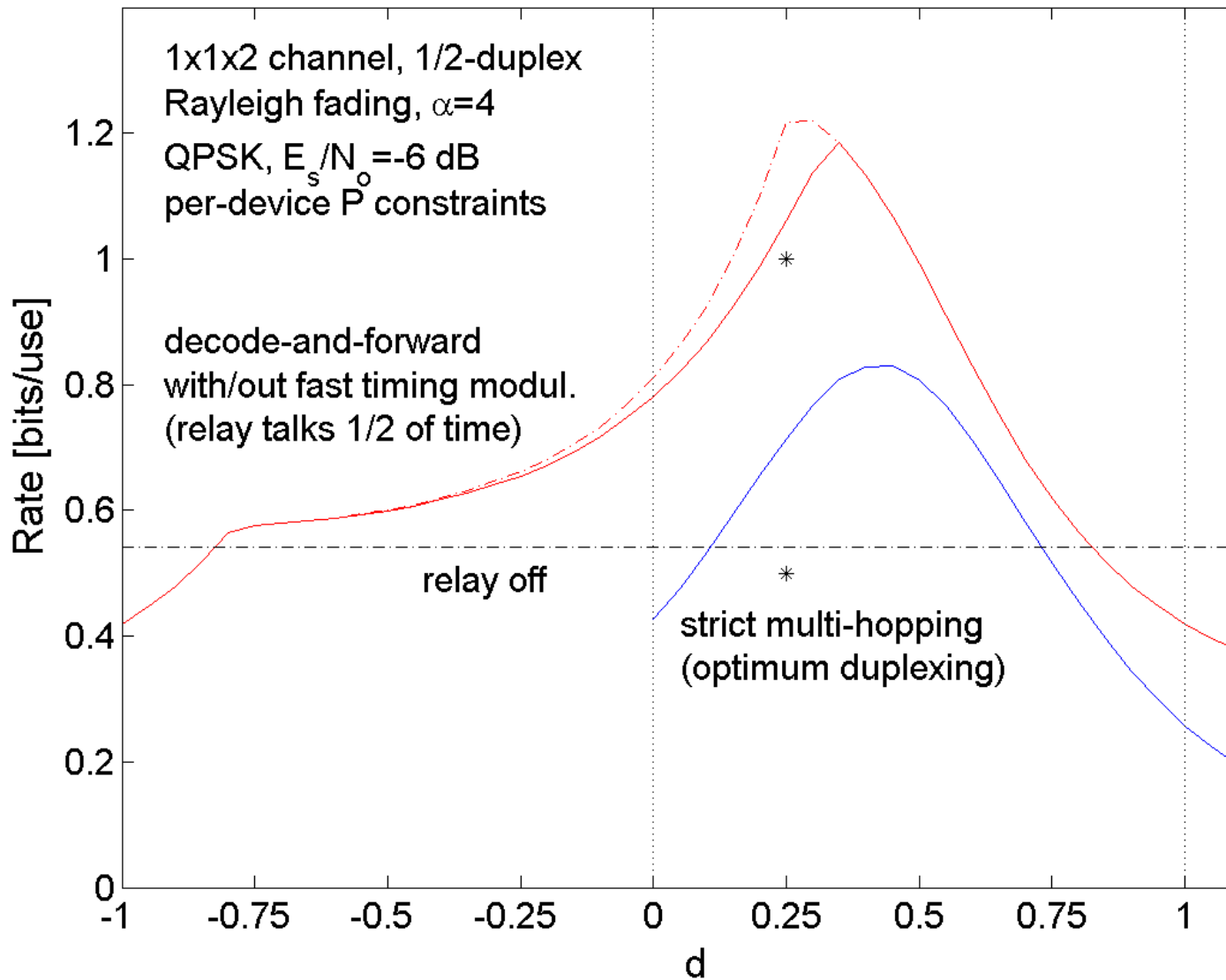
- Geometry:



- Parameters:

- Source and relay have 1 antenna, destin. has 2 antennas
- half-duplex relay
- Rayleigh fading, link gains known at the receivers only
- Attenuation exponent  $a=4$ , QPSK symbols
- $E_S/N_0 = -6$  dB, per-symbol & device power constraints,  $P_1 = P_2$
- Fast timing modulation with  $\Pr[M_2=0] = \Pr[M_2=1] = 1/2$

## Achievable Rates for a Half-Duplex Relay Channel



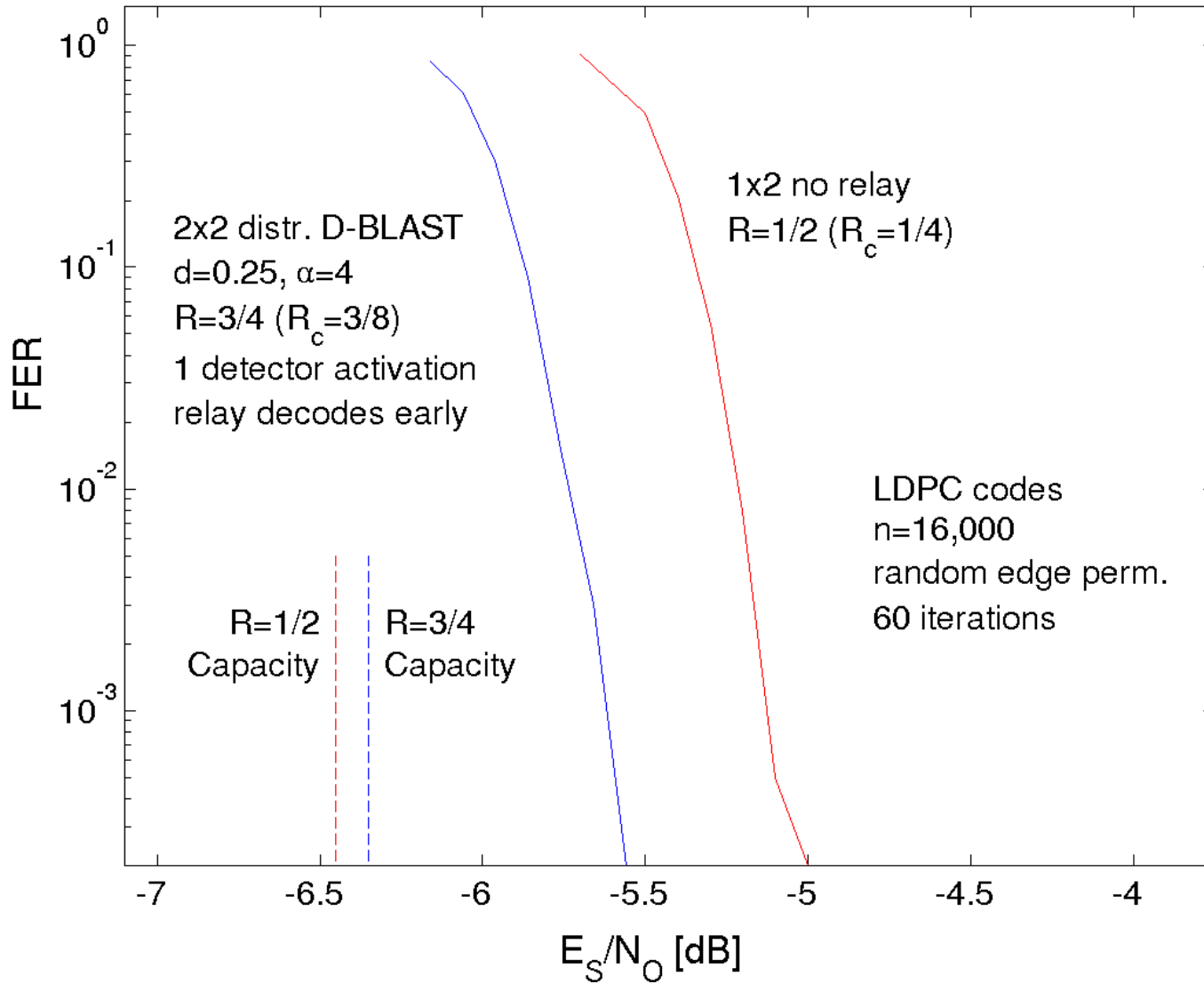
Note:  
- DF with timing mod. and optim. duplexing gives capacity if  $d$  is near zero



# Code Design Example

- Consider the above geometry with  $d=0.25$
- Go for  $R=1/2$  without relay and  $R=1$  with relay
- Use two LDPC codes designed for AWGN channels. Length  $n=16,000$  and design rates
  - $R_c=1/4$  for S? D link (code spans two blocks)  
(EXIT threshold  $E_b/N_0$  at -0.4 dB, capacity at -0.72 dB)
  - $R_c=3/8$  for S? D and SR? D links  
(EXIT threshold  $E_b/N_0$  at 0.1 dB, capacity at -0.35 dB)
- Receivers: 60 iterations
- Relay decodes after 4000 symbols (low error rate)
- Expect similar error rates for other two decoders

# Frame Error Rates for a Half-Duplex Relay Channel with Rayleigh Fading



Notes:

- within 1.3 dB of capacity at FER of  $10^{-3}$
- get closer by making  $n$  larger

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# Comparisons

- Some advantages over
  - multi-hopping
  - distributed space-time codes (various papers, 2001-present)
  - distributed V-BLAST (Augustín et al, Barbarossa et al (2004))

1) One can approach capacity (and outage capacity)

2) One has (almost) **flat** detector EXIT curves.

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# Summary

- Information-theoretic models and insights:
  - Let one understand basic limitations of **wireline** and **wireless** relaying (in small networks for now) in a unified way
  - Lead to new coding methods (e.g. timing modulation) that improve established methods (e.g., routing, network coding, multi-hopping)
  - Let one put some important methods for **relaying** (DF) and **multi-antenna transmission** in a common framework
  - Lead to new relaying methods that can approach capacity